An Introduction to Control Theory

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Purpose of the tutorial

• I do not know why you are locked up with me for a 4h tutorial in control theory! I only have some hints.

- At least, I know how the story started...
- The purposes of the tutorial are to:
 - 1. Give a short introduction to control theory.
 - 2. Show that some connections exist between control theory and commutative algebra (Lombardi, Coquand, Quitte...):

Fractional ideals, lattices, projective/stably free/free modules, Prüfer/Bézout domains, coherent rings, projective free rings, stable range, minimal generating systems...

Plan of the tutorial

- The plan of the tutorial is:
 - 1. Single-input single-ouput systems:

An introduction to the fractional ideal approach to stabilization problems

2. Multi-input multi-ouput systems:

An introduction to the lattice approach to stabilization problems

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Control theory

- Control theory can be divided into 3 main steps:
 - 1. **Modeling problems**: find a correct mathematical model for a real system coming from mechanics, electrical engineering, mathematical physics, biology...
 - 2. **Analysis problems**: analysis of the properties of the system (controllability, observability, stabilizability...).



3. **Synthesis problems**: construction of a feedback controller which stabilizes and optimizes the performances of the closed-loop system, study the robustness issues...



History & assumptions

- History of control theory:
 - 1. Prehistory: Watt (1769), Maxwell (1868), Lyapunov (1907),
 - 2. Frequency-domain approach (Black, Nyquist, Bode, 1930-40),
 - Time-domain approach (Bellman, Pontryagin, Kalman, 1957-60): state-space systems, controllability, observability, optimal control, Kalman filter...
 - 4. Robust control (Zames, Desoer, Francis, Doyle, 1980-90), μ -synthesis, Linear Matrix Inequalities (LMIs)...
 - 5. Future?
- We shall only study time-invariant linear systems defined by:
 - 1. ordinary or partial differential equations,
 - 2. differential time-delay equations.
- We shall focus on

synthesis problems within a frequency-domain approach.

An introduction to the fractional ideal approach to stabilization problems

- 1. Linear control systems
- 2. Laplace transform
- 3. Transfer function
- 4. Signal spaces and algebras
- 5. Stability
- 6. Fractional representation approach
- 7. Analysis problems
- 8. Synthesis problems
- 9. Theory of fractional ideals
- 10. NSC for internal/strong/robust stabilizability
- 11. Parametrization of all stabilizing controllers

Linear control systems

1. Finite-dimensional linear systems:

 $\dot{z}(t) = z(t) + u(t), \quad z(0) = 0, \quad y(t) = z(t).$

2. Infinite-dimensional linear systems:

2.1 Differential time-delay equations: $h \in \mathbb{R}_+$, i.e., $h \ge 0$,

$$\left\{ egin{array}{ll} \dot{z}(t) = z(t) + u(t), \ \ z(0) = 0, \ y(t) = \left\{ egin{array}{ll} 0, & 0 \leq t < h, \ z(t-h), & t \geq h. \end{array}
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2.2 Partial differential equations:

$$\begin{cases} \frac{\partial^2 z}{\partial t^2}(x,t) - a^2 \frac{\partial^2 z}{\partial x^2}(x,t) = 0, & x \in [0,l], \\ z(x,0) = 0, & \frac{\partial z}{\partial t}(x,0) = 0, \\ z(0,t) = u(t), & z(l,t) = 0, \\ y(t) = z(\overline{x},t), & \overline{x} \in [0,l]. \end{cases}$$

Laplace transform

$$L_1(\mathbb{R}_+) = \{ f \in \mathbb{R}_+ \to \mathbb{R} \mid \| f \|_1 = \int_0^{+\infty} |f(t)| \, dt < +\infty \}.$$

• Definition: Let $f \in \mathbb{R}_+ = [0, +\infty[\rightarrow \mathbb{R}]$ be a function such that:

$$\exists \alpha \in \mathbb{R} : e^{-\alpha t} f \in L_1(\mathbb{R}_+).$$

Then, the **Laplace transform** of *f* is defined by:

$$\mathcal{L}(f)(s) = \int_0^{+\infty} e^{-st} f(t) dt, \quad \forall s \in \mathbb{C}_{\alpha} = \{s \in \mathbb{C} \mid \operatorname{Re} s > \alpha\}.$$

- Notation: We also denote $\mathcal{L}(f)$ by \hat{f} .
- Example: $Y = \begin{cases} 1 & t > 0, \\ 0 & t \le 0 \end{cases}$, $\mathcal{L}(Y) = \frac{1}{s}$, $\mathcal{L}(\delta) = 1$, $\mathcal{L}(t^n Y) = \frac{(n+1)!}{s^{n+1}}$, $\mathcal{L}(t^n e^{-\lambda t} Y) = \frac{(n+1)!}{(s+\lambda)^{n+1}}$, $\mathcal{L}(e^{-\lambda t} \cos(\omega t) Y) = \frac{(s+\lambda)}{(s+\lambda)^2 + \omega^2}$.

Properties

• **Theorem**: If *f* is **Laplace transformable**, then we have:

1. \widehat{f} is analytic and bounded in $\mathbb{C}_{\alpha} = \{ s \in \mathbb{C} \mid \operatorname{Re} s > \alpha \}.$

- 2. If g is a Laplace transformable function such that $\widehat{f}(s) = \widehat{g}(s)$ in \mathbb{C}_{α} , for some $\alpha \in \mathbb{R}$, then f = g.
- 3. If $(f \star g)(t) = \int_0^t f(t \tau) g(\tau) d\tau$, $t \ge 0$, then:

$$\widehat{f \star g} = \widehat{f} \ \widehat{g}.$$

4. If
$$g(t) = \begin{cases} f(t-h), & t \ge h, \\ 0, & 0 < t < h, \end{cases}$$
, then $\widehat{g}(s) = e^{-hs} \widehat{f}(s)$.

5. If f is n times differentiable for t > 0 and $f^{(1)}, \ldots, f^{(n)}$ are Laplace transformable, then:

$$\widehat{f^{(n)}}(s) = s^n \,\widehat{f}(s) - \sum_{i=0}^{n-1} s^{n-i-1} \, f^{(i)}(0_+).$$

Transfer functions

• Ordinary differential equation:

$$\dot{z}(t)=z(t)+u(t), \ \ z(0)=0 \quad \Rightarrow \widehat{z}(s)=rac{1}{(s-1)}\,\widehat{u}(s).$$

• Differential time-delay equation:

$$\begin{cases} \dot{z}(t) = z(t) + u(t), \ x(0) = 0, \\ y(t) = \begin{cases} 0, & 0 \le t < h, \\ z(t-h), & t \ge h, \end{cases} \Rightarrow \widehat{y}(s) = \frac{e^{-hs}}{(s-1)} \widehat{u}(s). \end{cases}$$

• Wave equation:

$$\begin{cases} \frac{\partial^2 z}{\partial t^2}(x,t) - a^2 \frac{\partial^2 z}{\partial x^2}(x,t) = 0, \\ z(x,0) = 0, \quad \frac{\partial z}{\partial t}(x,0) = 0, \quad \Rightarrow \hat{y}(s) = \frac{\left(e^{-\frac{\overline{x}}{a}s} - e^{-\frac{(2l-\overline{x})s}{a}}\right)}{\left(1 - e^{-\frac{2a}{l}s}\right)} \, \hat{u}(s). \\ z(0,t) = u(t), \quad z(l,t) = 0, \\ y(t) = z(\overline{x},t), \end{cases}$$

Explicit computations

$$\begin{cases} \frac{\partial^2 z}{\partial t^2}(x,t) - a^2 \frac{\partial^2 z}{\partial x^2}(x,t) = 0, \\ z(x,0) = 0, \quad \frac{\partial z}{\partial t}(x,0) = 0, \\ z(0,t) = u(t), \quad z(l,t) = 0, \quad y(t) = z(\overline{x},t), \end{cases}$$
$$\frac{\partial^2 z}{\partial t^2}(x,t) - a^2 \frac{\partial^2 z}{\partial x^2}(x,t) = 0 \quad \Rightarrow \frac{d^2 \widehat{z}(x,s)}{dx^2} - \frac{s^2}{a^2} \widehat{z}(x,s) = 0, \\ \Rightarrow \widehat{z}(x,s) = A(s) e^{-\frac{s}{a}x} + B(s) e^{\frac{s}{a}x}. \end{cases}$$

$$\begin{cases} \widehat{z}(0,s) = \widehat{u}(s), \\ \widehat{z}(l,s) = 0, \end{cases} \Rightarrow \begin{cases} A(s) = \frac{1}{\left(1 - e^{-\frac{2a}{l}s}\right)} \widehat{u}(s), \\ B(s) = -\frac{e^{-\frac{2a}{l}s}}{\left(1 - e^{-\frac{2a}{l}s}\right)} \widehat{u}(s), \end{cases}$$
$$\Rightarrow \widehat{y}(s) = \widehat{z}(\overline{x}, s) = \frac{\left(e^{-\frac{\overline{x}}{a}s} - e^{-\frac{(2l-\overline{x})s}{a}}\right)}{\left(1 - e^{-\frac{2a}{l}s}\right)} \widehat{u}(s). \end{cases}$$

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Transfer functions

• Heat equation:

$$\begin{cases} \frac{\partial z}{\partial t}(x,t) - \lambda^2 \frac{\partial^2 z}{\partial x^2}(x,t) = 0, \\ z(x,0) = 0, \qquad \Rightarrow \widehat{y}(s) = \frac{\left(e^{\lambda (l-\overline{x})\sqrt{s}} - e^{-\lambda (l-\overline{x})\sqrt{s}}\right)}{\left(e^{\lambda l\sqrt{s}} - e^{-\lambda l\sqrt{s}}\right)} \widehat{u}(s). \\ z(0,t) = u(t), \quad z(l,t) = 0, \\ y(t) = z(\overline{x},t), \end{cases}$$

• Telegraph equation:

$$\begin{cases} \frac{\partial^2 z}{\partial t^2}(x,t) - a^2 \frac{\partial^2 z}{\partial x^2}(x,t) - k \, z(x,t) = 0, \\ z(x,0) = 0, \quad \frac{\partial z}{\partial t}(x,0) = 0, \qquad \Rightarrow \hat{y}(s) = e^{\frac{-\sqrt{s^2 - k}}{a} \,\overline{x}} \, \hat{u}(s). \\ z(0,t) = u(t), \quad \lim_{x \to +\infty} z(x,t) = 0, \\ y(t) = z(\overline{x},t), \end{cases}$$

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Transfer functions

• Electric transmission line:

$$\begin{cases} \frac{\partial V}{\partial x}(x,t) + L \frac{\partial I}{\partial t}(x,t) + R I(x,t) = 0, \\ \frac{\partial I}{\partial x}(x,t) + C \frac{\partial V}{\partial t}(x,t) + G V(x,t) = 0, \\ V(x,0) = 0, \quad I(x,0) = 0, \\ V(0,t) = u(t), \quad \lim_{x \to +\infty} V(x,t) = 0, \\ V(\overline{x},t) = y_1(t), \quad I(\overline{x},t) = y_2(t), \end{cases}$$
$$\Rightarrow \begin{cases} \hat{y}_1(s) = e^{-\sqrt{(Ls+R)(Cs+G)}\overline{x}} \hat{u}(s), \\ \hat{y}_2(s) = \sqrt{\frac{Cs+G}{Ls+R}} e^{-\sqrt{(Ls+R)(Cs+G)}\overline{x}} \hat{u}(s). \end{cases}$$

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Kernel representation (convolution)

• Inverse Laplace transform:

$$f(t) = \mathcal{L}^{-1}\left(\widehat{f}\right)(t) = \frac{1}{2\pi i} \int_{a-i\infty}^{a+i\infty} e^{st} \widehat{f}(s) \, ds = f(t), \ a > \alpha, \ t > 0.$$

• Ordinary differential equation:

$$\widehat{y}(s) = rac{1}{(s-1)} \widehat{u}(s) \Rightarrow y = \left(\mathcal{L}^{-1}\left(rac{1}{s-1}
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ight) \star u,$$

 $\Rightarrow y(t) = \int_0^t e^{t-\tau} u(\tau) d\tau, \quad t \ge 0.$

• Differential time-delay equation:

$$\widehat{y}(s) = \frac{e^{-hs}}{(s-1)} \widehat{u}(s) \Rightarrow y = \left(\mathcal{L}^{-1}\left(\frac{e^{-hs}}{s-1}\right)\right) \star u,$$

$$\Rightarrow y(t) = \int_0^{t-h} e^{t-h-\tau} u(\tau) d\tau, \quad t \ge h, \quad 0 \quad \text{else.}$$

Kernel representation (convolution)

• Wave equation:

 \Rightarrow

$$\widehat{y}(s) = \frac{\left(e^{-\frac{\overline{x}}{a}s} - e^{-\frac{(2l-\overline{x})s}{a}}\right)}{\left(1 - e^{-\frac{2a}{l}s}\right)} \widehat{u}(s)$$

$$\Rightarrow \widehat{y}(s) = \left(e^{-\frac{s}{a}\overline{x}} \sum_{n=0}^{+\infty} e^{-\frac{2ns}{a}I} - e^{-\frac{s}{a}(2I-\overline{x})} \sum_{n=0}^{+\infty} e^{-\frac{2ns}{a}I}\right) \widehat{u}(s),$$
$$\Rightarrow y(\overline{x}, t) = \sum_{n=0}^{+\infty} u\left(t - \frac{2nI+\overline{x}}{a}\right) - \sum_{n=1}^{+\infty} u\left(t - \frac{2nI-\overline{x}}{a}\right).$$

• Telegraph equation: $k = \beta^2 > 0$.

$$\widehat{y}(s) = e^{\frac{-\sqrt{s^2 - \beta^2}}{a}\overline{x}}\widehat{u}(s)$$

$$\Rightarrow y(\overline{x}, t) = u\left(t - \frac{\overline{x}}{a}\right) + \beta\left(\frac{\overline{x}}{a}\right) \int_{\frac{\overline{x}}{a}}^{t} u(t - \tau) \frac{h\left(\beta\sqrt{\tau^2 - \left(\frac{\overline{x}}{a}\right)^2}\right)}{\sqrt{\tau^2 - \left(\frac{\overline{x}}{a}\right)^2}} d\tau.$$

Signal spaces

- Let us define the right half plane $\mathbb{C}_+ = \{s \in \mathbb{C} \mid \operatorname{Re} s > 0\}.$
- The Hardy algebra $H_{\infty}(\mathbb{C}_+)$ is defined by:

 $H_{\infty}(\mathbb{C}_{+}) = \{ \text{analytic functions } f \text{ in } \mathbb{C}_{+} \mid \| f \|_{\infty} = \sup_{s \in \mathbb{C}_{+}} |f(s)| < +\infty \}.$

 $H_{\infty}(\mathbb{C}_+)$ is a commutative **Banach algebra**.

• The Hardy vector-space $H_2(\mathbb{C}_+)$ is defined by:

 $\begin{aligned} H_2(\mathbb{C}_+) &= \{ \text{ analytic functions } f \text{ in } \mathbb{C}_+ \mid \\ &\parallel f \parallel_2 = \sup_{x \in \mathbb{R}_+} \left(\int_{-\infty}^{+\infty} |f(x + iy)|^2 dy \right)^{1/2} < +\infty \} \end{aligned}$

 $H_2(\mathbb{C}_+)$ is a **Hilbert space** and $H_2(\mathbb{C}_+) = \mathcal{L}(L_2(\mathbb{R}_+))$, where:

$$L_2(\mathbb{R}_+) = \{g: \mathbb{R}_+ o \mathbb{R} \mid \left(\int_0^{+\infty} |g(t)|^2 dt\right)^{1/2} < +\infty\}.$$

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$L_2(\mathbb{R}_+) - L_2(\mathbb{R}_+)$ -stability

• Theorem:

1. $\forall a, b \in H_{\infty}(\mathbb{C}_+), \forall f, g \in H_2(\mathbb{C}_+): af + bg \in H_2(\mathbb{C}_+).$

2. The linear operator

$$\begin{array}{rcl} \Lambda: H_2(\mathbb{C}_+) & \longrightarrow & H_2(\mathbb{C}_+), \\ & u & \longmapsto & h \, u, \end{array}$$

is **bounded**, i.e.:

 $\operatorname{dom}(\Lambda) = \{ u \in H_2(\mathbb{C}_+) \mid \Lambda(u) \in H_2(\mathbb{C}_+) \} = H_2(\mathbb{C}_+),$

iff $h \in H_{\infty}(\mathbb{C}_+)$. Then, we have:

$$\| \Lambda \|_{\mathcal{L}} (H_2(\mathbb{C}_+), H_2(\mathbb{C}_+)) = \sup_{0 \neq u \in H_2(\mathbb{C}_+)} \frac{\| h u \|_2}{\| u \|_2} = \| h \|_{\infty}.$$

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$$p = \frac{1}{s-1} \notin H_{\infty}(\mathbb{C}_{+})$$
 as p has a pole at $1 \in \mathbb{C}_{+}$,
 $\Rightarrow \Lambda : H_{2}(\mathbb{C}_{+}) \longrightarrow H_{2}(\mathbb{C}_{+}),$
 $\hat{u} \longmapsto \hat{y} = \frac{1}{(s-1)}\hat{u},$ is unbounded,
 $\Rightarrow \lambda : L_{2}(\mathbb{R}_{+}) \longrightarrow L_{2}(\mathbb{R}_{+}),$
 $u \longmapsto y = e^{t} Y \star u,$ is unbounded,
 $\dim \Lambda = \left(\frac{s-1}{s+1}\right) H_{2}(\mathbb{C}_{+}), \quad \dim \lambda = (\delta - 2e^{-t}Y) \star L_{2}(\mathbb{R}_{+}).$
• $u = e^{-t}Y \in L_{2}(\mathbb{R}_{+}): \quad || u ||_{2} = \frac{1}{\sqrt{2}}, \quad \hat{u} = \frac{1}{s+1} \in H_{2}(\mathbb{C}_{+}),$
 $\Rightarrow \hat{y} = \frac{1}{s^{2}-1} \notin H_{2}(\mathbb{C}_{+}), \quad \frac{1}{s^{2}-1} = \mathcal{L}((\operatorname{sh} t) Y).$
• $y(t) = \int_{0}^{t} e^{t-\tau} e^{-\tau} d\tau = (\operatorname{sh} t)Y \notin L_{2}(\mathbb{R}_{+}).$

•
$$p = \frac{e^{-hs}}{s-1} \notin H_{\infty}(\mathbb{C}_{+})$$
 as p has a pole at $1 \in \mathbb{C}_{+}$,
 $\Rightarrow \Lambda : H_{2}(\mathbb{C}_{+}) \longrightarrow H_{2}(\mathbb{C}_{+}),$
 $\widehat{u} \longmapsto \widehat{y} = \frac{e^{-hs}}{(s-1)}\widehat{u},$ is unbounded,
 $\Rightarrow \lambda : L_{2}(\mathbb{R}_{+}) \longrightarrow L_{2}(\mathbb{R}_{+}),$
 $u \longmapsto y = e^{t-h}Y \star u,$ is unbounded.
• $u = e^{-t}Y \in L_{2}(\mathbb{R}_{+})$: $||u||_{2} = \frac{1}{\sqrt{2}}, \quad \widehat{u} = \frac{1}{s+1} \in H_{2}(\mathbb{C}_{+}),$
 $\Rightarrow \widehat{y} = \frac{e^{-hs}}{s^{2}-1} \notin H_{2}(\mathbb{C}_{+}), \quad \frac{e^{-hs}}{s^{2}-1} = \mathcal{L}((\operatorname{sh}(t-h))Y).$
• $y(t) = \int_{0}^{t-h} e^{t-h-\tau} e^{-\tau} d\tau = (\operatorname{sh}(t-h))Y \notin L_{2}(\mathbb{R}_{+}).$

• The transfer function $p = \frac{e^{-\frac{\overline{x}}{a}s} - e^{-\frac{(2l-\overline{x})}{a}s}}{1 - e^{-\frac{2l}{d}s}}$ is such that $\| p \|_{\infty} = +\infty$ as p has poles at $s_k = \frac{l}{a}\pi k i, \ k \in \mathbb{Z}$

$$\Rightarrow p \text{ is not } H_2(\mathbb{C}_+) - H_2(\mathbb{C}_+) \text{-stable}.$$

- The transfer function $p = \frac{1}{s} \notin H_{\infty}(\mathbb{C}_+)$ as $|| p ||_{\infty} = +\infty$.
 - $\Rightarrow \Lambda : H_2(\mathbb{C}_+) \longrightarrow H_2(\mathbb{C}_+),$ $\widehat{u} \longmapsto \widehat{y} = \frac{1}{s} \widehat{u},$ is unbounded,

$$\Rightarrow \lambda : L_2(\mathbb{R}_+) \longrightarrow L_2(\mathbb{R}_+),$$

$$u \longmapsto y(t) = \int_0^t u(\tau) d\tau,$$
 is unbounded.

 $u(t) = Y/(t+1) \in L_2(\mathbb{R}_+), \parallel u \parallel_2 = 1, y = \ln(1+t) \notin L_2(\mathbb{R}_+).$

Signal spaces

•
$$L_1(\mathbb{R}_+) = \{ f : [0, +\infty[\to \mathbb{R} \mid \| f \|_1 = \int_0^{+\infty} |f(t)| dt < +\infty \},$$

$$l_1(\mathbb{Z}_+) = \{ a : \mathbb{Z}_+ = \{0, 1, \ldots\} o \mathbb{R} \mid \| (a_i)_{i \in \mathbb{Z}_+} \|_1 = \sum_{i=0}^{+\infty} |a_i| < +\infty \}.$$

• Definition: The Wiener algebra \mathcal{A} is defined by:

$$\mathcal{A} = \{ f = g + \sum_{i=0}^{+\infty} a_i \, \delta_{(t-h_i)} \mid g \in L_1(\mathbb{R}_+), \ (a_i)_{i \in \mathbb{Z}_+} \in I_1(\mathbb{Z}_+), \\ 0 = h_0 \le h_1 \le h_2 \le \ldots \}.$$

• \mathcal{A} is a commutative **Banach algebra** w.r.t.:

$$\parallel f \parallel_{\mathcal{A}} = \parallel g \parallel_1 + \parallel (a_i)_{i \in \mathbb{Z}_+} \parallel_1$$
.

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• $\widehat{\mathcal{A}} = \{\mathcal{L}(f) \mid f \in \mathcal{A}\}, \quad \| \widehat{f} \|_{\widehat{\mathcal{A}}} = \| f \|_{\mathcal{A}}.$

$L_\infty(\mathbb{R}_+) - L_\infty(\mathbb{R}_+)$ -stability

- Theorem: Let $p \in [1, +\infty[$.
 - 1. $\forall a, b \in \mathcal{A}, \quad \forall f, g \in L_p(\mathbb{R}_+): \quad a \star f + b \star g \in L_p(\mathbb{R}_+).$
 - 2. The linear operator

$$\begin{array}{rcl} \Lambda: L_{\infty}(\mathbb{R}_{+}) & \longrightarrow & L_{\infty}(\mathbb{R}_{+}), \\ & u & \longmapsto & h \star u, \end{array}$$

is **bounded**, i.e., $\operatorname{dom} \Lambda = L_{\infty}(\mathbb{R}_+)$, iff $\widehat{h} \in \widehat{\mathcal{A}}$ and:

$$\|\wedge\|_{\mathcal{L}(L_{\infty}(\mathbb{R}_{+}),L_{\infty}(\mathbb{R}_{+}))} = \sup_{0\neq u\in L_{\infty}(\mathbb{R}_{+})} \frac{\|h\star u\|_{\infty}}{\|u\|_{\infty}} = \|\hat{h}\|_{\widehat{\mathcal{A}}}.$$

3. $\hat{f} \in \widehat{\mathcal{A}}$ is analytic and bounded in $\overline{\mathbb{C}_+} = \{s \in \mathbb{C} \mid \operatorname{Re} s \ge 0\}$ and continuous on $i \mathbb{R}$:

$$\|\widehat{f}\|_{\infty} \leq \|\widehat{f}\|_{\widehat{\mathcal{A}}}, \quad \widehat{\mathcal{A}} \subset H_{\infty}(\mathbb{C}_{+}) \quad (e^{-\frac{1}{s}} \in H_{\infty}(\mathbb{C}_{+}) \setminus \widehat{\mathcal{A}}).$$

4. BIBO-stability $\Rightarrow L_p(\mathbb{R}_+) - L_p(\mathbb{R}_+)$ -stability.

•
$$\frac{1}{s-1} \notin H_{\infty}(\mathbb{C}_+) \Rightarrow \frac{1}{s-1} \notin \widehat{\mathcal{A}} \quad (\widehat{\mathcal{A}} \subset H_{\infty}(\mathbb{C}_+)).$$

Let $e^{-t} Y \in L_{\infty}(\mathbb{R}_+)$, $\| e^{-t} Y \|_{\infty} = 1$. Then, we have:

$$y(t) = \int_0^t e^{t- au} e^{- au} d au = (\operatorname{sh} t) Y \notin L_\infty(\mathbb{R}_+).$$

•
$$\frac{e^{-hs}}{s-1} \notin H_{\infty}(\mathbb{C}_+) \Rightarrow \frac{e^{-hs}}{s-1} \notin \widehat{\mathcal{A}}$$
. Let us take $e^{-t} Y \in L_{\infty}(\mathbb{R}_+)$

$$\Rightarrow y(t) = \int_0^{t-h} e^{t-h-\tau} e^{-\tau} d\tau = (\operatorname{sh}(t-h)) Y \notin L_{\infty}(\mathbb{R}_+).$$

•
$$p = \frac{e^{-\frac{\overline{X}}{a}s} - e^{-\frac{(2I-\overline{X})}{a}s}}{1 - e^{-\frac{2a}{I}s}} \notin H_{\infty}(\mathbb{C}_{+}) \Rightarrow p \in \widetilde{\mathcal{A}}, \text{ i.e.:}$$

$$h = \sum_{n=0}^{+\infty} \delta_{\left(t - \frac{2nl + \overline{x}}{a}\right)} - \sum_{n=1}^{+\infty} \delta_{\left(t - \frac{2nl - \overline{x}}{a}\right)} \notin \mathcal{A}.$$

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 \Rightarrow The 3 plants are **not BIBO stable**.

Control theory

• Let the **open-loop** $\widehat{u} \mapsto \widehat{y} = p \widehat{u}$ be **unstable**.

Control theory: stabilization by feedback.

• Is it possible to find a controller *c* such that the **closed-loop is** stable $\forall \ \hat{u}_1, \ \hat{u}_2 \in H_2(\mathbb{C}_+) \ (\forall \ u_1, \ u_1 \in L_{\infty}(\mathbb{R}_+))$?



• Can we parametrize the set of stabilizing controllers of p?

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• Is it possible to find **robust/optimal controllers** *c* of *p*?

The fractional representation of plants

• (Zames) The set of transfer functions has the structure of an algebra (parallel +, serie \circ , proportional feedback . by \mathbb{R}).

• (Vidyasagar) Let A be an algebra of stable transfer functions with a structure of an integral domain $(a \ b = 0, \ a \neq 0 \Rightarrow b = 0)$ and its the field of fractions:

$$K = Q(A) = \{p = n/d \mid 0 \neq d, n \in A\}.$$

K represents the class of systems

 $p \in A \Rightarrow p$ is stable, $p \in (K \setminus A) \Rightarrow p$ is unstable.

• (Zames) The algebra *A* of stable transfer functions has to be a normed algebra so that we can consider the errors in the modelization & approximation of the real plant by a model

(e.g., A is a **Banach algebra**: $\| a b \|_{A} \le \| a \|_{A} \| b \|_{A}, \| 1 \|_{A} = 1$).

Quotation

"... As soon as I read this, my immediate reaction was 'What is so difficult about handling that case? All one has to do is to write the unstable part as a ratio of two stable rational functions!' Without exaggeration, I can say that the idea occurred to me within no more than 10 min. So there it is the best idea I have had in my entire research career, and it took less than 10 min.

All the thousands of hours I have spent thinking about problems in control theory since have not resulted in any ideas as good as this one. I don't think I know what the 'moral of this story' really is !'',

"... It turns out that this seemingly simple stratagem leads to conceptually simple and computationally tractable solutions to many important and interesting problems."

M. Vidyasagar, "A brief history of the graph topology", *European* J. of Control, 2 (1996), 80-87.

• Let $RH_{\infty} = \mathbb{R}(s) \cap H_{\infty}(\mathbb{C}_+)$ be the algebra of exponentially-stable finite-dimensional plants, i.e.:

 $RH_{\infty} = \left\{ n/d \in \mathbb{R}(s) \ | \ \deg n \leq \deg d, \ d(\overline{s}) = 0 \Rightarrow \operatorname{Re} \overline{s} < 0 \right\}.$

$$p = \frac{1}{s-1} = \frac{\left(\frac{1}{s+1}\right)}{\left(\frac{s-1}{s+1}\right)}, \quad \frac{1}{s+1}, \quad \frac{s-1}{s+1} \in RH_{\infty} \Rightarrow p \in Q(RH_{\infty}).$$

• $\hat{\mathcal{A}}$: algebra of BIBO-stable ∞ -dimensional plants:

$$p = \frac{e^{-hs}}{s-1} = \frac{\left(\frac{e^{-hs}}{s+1}\right)}{\left(\frac{s-1}{s+1}\right)}, \quad \frac{e^{-hs}}{s+1}, \quad \frac{s-1}{s+1} \in \hat{\mathcal{A}} \Rightarrow p \in Q(\hat{\mathcal{A}}).$$

• $H_{\infty}(\mathbb{C}_+)$: algebra of $L_2(\mathbb{R}_+)$ -stable ∞ -dimensional plants:

$$p = \frac{(1+e^{-2s})}{(1-e^{-2s})} \in Q(H_{\infty}(\mathbb{C}_{+})): \quad 1+e^{-2s}, \ 1-e^{-2s} \in H_{\infty}(\mathbb{C}_{+}).$$

(Weakly) coprime factorization

• Let A be an algebra of **stable transfer functions** and:

$$K = Q(A) = \{n/d, 0 \neq d, n \in A\}.$$

- **Definition**: A transfer function $p \in K$ is said to admit a **weakly** coprime factorization if:
 - $\exists \ 0 \neq d, \ n \in A: \quad p = n/d, \quad \forall \ k \in K: \ k \ n, \ k \ d \in A \Rightarrow k \in A.$
- **Definition**: A transfer function $p \in K$ is said to admit a **coprime factorization over** A if:

$$\exists 0 \neq d, n, x, y \in A: p = n/d, dx - ny = 1.$$

• A coprime factorization is a weakly coprime factorization:

$$k \in K$$
: $k n, k d \in A \Rightarrow k = (k d) x - (k n) y \in A$

• **Example**: Let $A = RH_{\infty}$ and $p = \frac{1}{(s-1)} \in \mathbb{R}(s)$. Then,

$$p = \frac{n}{d}, \ n = \frac{1}{(s+1)(s+2)}, \ d = \frac{(s-1)}{(s+1)(s+2)} \in A,$$

is not a weakly coprime facorization as:

$$(s+2) \in Q(A) = \mathbb{R}(s), \quad (s+2) \notin A, \quad \left\{ egin{array}{ll} (s+2) \ n = rac{1}{(s+1)} \in A, \ (s+2) \ d = rac{(s-1)}{(s+1)} \in A. \end{array}
ight.$$

• **Example**: Let $A = RH_{\infty}$ and $p = \frac{1}{(s-1)} \in \mathbb{R}(s)$. Then,

$$p=rac{n}{d},\quad n=rac{1}{(s+1)},\quad d=rac{(s-1)}{(s+1)}\in A,$$

is a **coprime factorization** of *p* as we have:

$$\frac{(s-1)}{(s+1)} - (-2)\frac{1}{(s+1)} = 1, \quad x = 1, \quad y = -2.$$

Internal stabilizability

- Let A be an algebra of stable transfer functions, K = Q(A).
- Let $p \in K$ be a **plant** and $c \in K$ a **controler**.
- The closed-loop system is defined by:

$$\left(\begin{array}{c}u_1\\u_2\end{array}\right)=\left(\begin{array}{cc}1&-p\\-c&1\end{array}\right)\left(\begin{array}{c}e_1\\e_2\end{array}\right),\quad\begin{cases}y_1=e_2-u_2,\\y_2=e_1-u_1.\end{cases}$$

• **Definition**: *c* **internally stabilizes** *p* if we have:

$$H(p,c) = \begin{pmatrix} 1 & -p \\ -c & 1 \end{pmatrix}^{-1} = \begin{pmatrix} \frac{1}{1-pc} & \frac{p}{1-pc} \\ \frac{c}{1-pc} & \frac{1}{1-pc} \end{pmatrix} \in A^{2\times 2}$$

 \Rightarrow *c* is then called a **stabilizing controler** of *p*.

• Example: $A = RH_{\infty}$, $K = Q(A) = \mathbb{R}(s)$.

$$\begin{cases} p = \frac{s}{(s-1)}, \\ c = -\frac{(s-1)}{(s+1)}, \end{cases} \Rightarrow \begin{cases} e_1 = \frac{(s+1)}{(2s+1)} u_1 + \frac{s(s+1)}{(2s+1)(s-1)} u_2, \\ e_2 = \frac{(-s+1)}{(2s+1)} u_1 + \frac{(s+1)}{(2s+1)} u_2. \end{cases}$$

 \Rightarrow *c* does not internally stabilize *p* because:

$$rac{s(s+1)}{(2s+1)(s-1)}
otin extsf{RH}_{\infty} \hspace{0.2cm} (extsf{pole in } 1 \in \mathbb{C}_{+}).$$

• Example: $A = RH_{\infty}$, $K = Q(A) = \mathbb{R}(s)$.

$$\begin{cases} p = \frac{s}{(s-1)}, \\ c = 2, \end{cases} \Rightarrow \begin{cases} e_1 = -\frac{(s-1)}{(s+1)} u_1 - \frac{s}{(s+1)} u_2, \\ e_2 = -2 \frac{(s-1)}{(s+1)} u_1 - \frac{(s-1)}{(s+1)} u_2. \end{cases}$$

 \Rightarrow *c* internally stabilizes the plant *p*.

Strong and simultaneous stabilizations

- Let A be an algebra of stable transfer functions, K = Q(A).
- Definition: $p \in K$ is strongly stabilizable if there exists a stable controller c, i.e., $c \in A$, which internally stabilizes p.
- **Definition**: The plants $p_1, \ldots, p_n \in K$ are **simultaneously stabilizable** if $\exists c \in K$ which internally stabilizes p_1, \ldots, p_n .
- Interests of the strong stabilization:

Safety, good ability to track reference inputs.

• Interests of the simultaneous stabilization:

The controller is designed to stabilize a family of plants, e.g.: operating conditions, failed modes, loss of sensors/actuators, changes of operating points.

• Example: Let $A = RH_{\infty}$. The plant $p = \frac{1}{(s-1)}$ is strongly stabilized by $c = -2 \in A$ as we have:

$$\frac{1}{1-p\,c} = \frac{(s-1)}{(s+1)}, \quad \frac{p}{1-p\,c} = \frac{1}{(s+1)}, \quad \frac{c}{1-p\,c} = -\frac{2\,(s-1)}{(s+1)}.$$

• Example: Let $A = H_{\infty}(\mathbb{C}_+)$. The plant $p = \frac{(1+e^{-2s})}{(1-e^{-2s})} \in K$ is strongly stabilized by c = -1 as we have:

$$\frac{1}{1-pc} = \frac{1-e^{-2s}}{2}, \ \frac{p}{1-pc} = \frac{1+e^{-2s}}{2}, \ \frac{c}{1-pc} = -\frac{1-e^{-2s}}{2}.$$

• **Example**: Let $A = RH_{\infty}$. The plants defined by

$$p_1 = rac{1}{(s+1)}, \quad p_2 = rac{2 \, s}{(s-1) \, (s+1)},$$

are simultaneously stabilized by $c = -2 \frac{(s+1)}{(s-1)}$.

Robust stabilizability

• Let A be a Banach algebra of stable transfer functions

(e.g.,
$$A = H_{\infty}(\mathbb{C}_+), \quad \widehat{\mathcal{A}}, \quad A(\mathbb{D}), \quad W_+).$$

- **Definition**: Let $c \in K = Q(A)$ be a stabilizing controller of $p \in K$. Then, *c* robustly stabilizes *p* if there exits $\epsilon > 0$ such that *c* internally stabilizes one of the family of plants:
 - 1. Additive perturbations:

$$B_1(p,\delta) = \{p + \delta \mid \forall \, \delta \in A, \parallel \delta \parallel_A < \epsilon \}.$$

2. Multiplicative perturbations:

 $B_2(\boldsymbol{p}, \delta) = \{ \boldsymbol{p}/(1 + \delta \boldsymbol{p}) \mid \forall \, \delta \in \boldsymbol{A}, \parallel \delta \parallel_{\boldsymbol{A}} < \epsilon \}.$

3. Relative additive perturbations:

$$B_3(p,\delta) = \{ p(1+\delta) \mid \forall \, \delta \in A, \parallel \delta \parallel_A < \epsilon \}.$$

4. Relative multiplicative perturbations:

$$B_4(p,\delta) = \{ p/(1+\delta) \, | \, \forall \, \delta \in \mathcal{A}, \, \| \, \delta \, \|_{\mathcal{A}} < \epsilon \}.$$

Theory of fractional ideals

- Let A be an integral domain and $K = \{n/d \mid 0 \neq d, n \in A\}$.
- Definition: A fractional ideal J of A is an A-submodule of K

$$(\forall a_1, a_2 \in A, \forall m_1, m_2 \in J: a_1 m_1 + a_2 m_2 \in J),$$

such that $\exists 0 \neq d \in A$ satisfying:

$$(d) J \triangleq \{a d \mid a \in J\} \subseteq A.$$

• **Example**: Let A be an algebra of stable transfer functions and $p \in K = Q(A)$ a **transfer function**. Then,

$$J = (1, p) \triangleq A + A p$$

is a **fractional ideal** of A as:

$$\exists 0 \neq d, n \in A : p = n/d \Rightarrow (d) J = A d + A n \subseteq A.$$

•
$$y = p \ u \Rightarrow (1, -p) (y \ u)^T = 0 \Rightarrow J = (1, -p) = (1, p) = (1, p)$$

Theory of fractional ideals

- **Definition**: A fractional ideal J of A is **integral** if $J \subseteq A$.
- **Example**: If $p \in A$, then J = (1, p) = A. Conversely,

$$J = (1, p) = (1) \Rightarrow \exists n \in A : p = n1 = n \in A.$$

 \Rightarrow the transfer function p is **stable** iff J = (1, p) = A.

• **Definition**: A fractional ideal J of A is **principal** if $\exists k \in K$:

$$J = (k) \triangleq A k = \{a k \mid a \in A\}.$$

• **Example**: J = (1, p) is principal iff there exists $0 \neq k \in K$ such that J = (k), i.e., iff there exist $0 \neq d$, n, x, $y \in A$ s.t.:

$$\begin{cases} 1 = d k, \\ p = n k, \\ k = x - y p \end{cases} \Leftrightarrow \begin{cases} k = 1/d, \\ p = n/d, \\ 1/d = x - y (n/d), \end{cases} \Leftrightarrow \begin{cases} p = n/d, \\ d x - n y = 1. \end{cases}$$

⇒ the transfer function p admits a coprime factorization p = n/d iff J = (1/d), i.e., J is principal.

- Let $A = H_{\infty}(\mathbb{C}_+)$ and $p = \frac{e^{-s}}{(s-1)} \in \mathcal{K} = Q(A)$.
- Let J = (1, p) be the fractional ideal of A defined by 1 and p.
- We have $J = \left(\frac{s+1}{s-1}\right)$ as we have:

$$\begin{cases} 1 = \left(\frac{s-1}{s+1}\right) \left(\frac{s+1}{s-1}\right), \\\\ \frac{e^{-s}}{(s-1)} = \left(\frac{e^{-s}}{s+1}\right) \left(\frac{s+1}{s-1}\right), \\\\ \frac{(s+1)}{(s-1)} = \left(1+2\left(\frac{1-e^{-(s-1)}}{s-1}\right)\right) + 2e\frac{e^{-s}}{(s-1)} \qquad (\star). \end{cases}$$

 $p = \frac{n}{d}$, $n = \frac{e^{-s}}{(s+1)}$, $d = \frac{(s-1)}{(s+1)}$, is a coprime factorization of p:

$$(\star) \Leftrightarrow \left(\frac{s-1}{s+1}\right) \left(1+2\left(\frac{1-e^{-(s-1)}}{s-1}\right)\right) - \left(\frac{e^{-s}}{s+1}\right) (-2e) = 1.$$

Theory of fractional ideals

• Proposition: Let $\mathcal{F}(A)$ be the set of non-zero fractional ideals of A and I, $J \in \mathcal{F}(A)$. Then, we have:

$$\begin{cases} I J = \{\sum_{\text{finite}} a_i b_i \mid a_i \in I, b_i \in J\} \in \mathcal{F}(A), \\ I : J = \{k \in K \mid (k) J \subseteq I\} \in \mathcal{F}(A). \end{cases}$$

• **Example**: Let $p \in K$ and J = (1, p). Then, we have

$$A: J = \{k \in K \mid k, k p \in A\} = \{d \in A \mid d p \in A\}$$

is called the ideal of the denominators of p.

p admits a weakly coprime factorization p = n/d iff:

$$\exists 0 \neq d, n \in A : (d, n) = \{k \in K \mid k d, k n \in A\} = A,$$

$$\Leftrightarrow A : ((d) (1, p)) = A \Leftrightarrow (A : J) : (d) = A$$

$$\Leftrightarrow (d^{-1}) (A : J) = A \Leftrightarrow A : J = (d).$$

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• Let A be the Banach algebra of analytic functions in the unit disc \mathbb{D} whose Taylor series converge absolutely, i.e.:

$$W_{+} = \left\{ f(z) = \sum_{i=0}^{+\infty} a_{i} z^{i} \mid \sum_{i=0}^{+\infty} |a_{i}| < +\infty \right\}.$$

- A is the algebra of the **BIBO-stable causal filters**.
- Let us consider the transfer function $p = e^{-\left(\frac{1+z}{1-z}\right)}$:

$$\left\{ egin{array}{l} n=(1-z)^3\,e^{-\left(rac{1+z}{1-z}
ight)}\in A,\ d=(1-z)^3\in A, \end{array}
ight. \Rightarrow p=n/d\in Q(A).$$

- Let us consider the **fractional ideal** J = (1, p) of A.
- The ideal $A : J = \{ d \in A \mid d p \in A \}$ is not finitely generated.

See R. Mortini & M. Von Renteln, "Ideals in Wiener algebra", J. Austral. Math. Soc., 46 (1989), 220-228.

 $\Rightarrow p$ does not admit a (weakly) coprime factorization.

• The disc algebra $A(\mathbb{D})$ is the Banach algebra of holomophic functions in the open unit disc $\mathbb{D} = \{z \in \mathbb{C} \mid |z| < 1\}$ which are continuous on the unit circle $\mathbb{T} = \{z \in \mathbb{C} \mid |z| = 1\}$.

• We have $n = (1-z) e^{-\left(rac{1+z}{1-z}
ight)} \in A$, $d = (1-z) \in A$,

$$\Rightarrow p = n/d = e^{-\left(rac{1+z}{1-z}
ight)} \in Q(A), \quad J = (1, p).$$

• $A : J = \{d \in A \mid d p \in A\} = \{d \in A \mid d(1) = 0\}$ is a maximal ideal of A which is **not finitely generated**.

See R. Mortini, "Finitely generated prime ideals in H^{∞} and $A(\mathbb{D})$ ", Math. Z., 191 (1986), 297-302.

 $\Rightarrow p$ does not admit a (weakly) coprime factorization and p is not internally stabilizable.

Theory of fractional ideals

• **Definition**: $J \in \mathcal{F}(A)$ is invertible if $\exists I \in \mathcal{F}(A)$:

I J = A.

• Proposition: If J is an invertible fractional ideal of A, then:

$$I = A : J = \{k \in K \mid (k) J \subseteq A\}.$$

• If J is an invertible fractional ideal of A, we then denote by:

$$I = J^{-1}$$

• **Proposition**: If *J* is invertible, then we have:

$$(J^{-1})^{-1} = J.$$

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Theory of fractional ideals

- Let $p \in K$ and J = (1, p). If J is **invertible**, then we have:
- $1 \in J(A:J) = (1, p) \left(\{ d \in A \mid d p \in A \} \right) = \{ \alpha + \beta p \mid \alpha, \beta \in A : J \}$ $\Leftrightarrow \exists a, b \in A : \begin{cases} a b p = 1, \\ a p \in A, b p \in A. \end{cases}$

If $a \neq 0$, then $c = b/a \in K$ satisfies:

$$H(p, c) = \begin{pmatrix} \frac{1}{1-pc} & \frac{p}{1-pc} \\ \frac{c}{1-pc} & \frac{1}{1-pc} \end{pmatrix} = \begin{pmatrix} a & ap \\ b & a \end{pmatrix} \in A^{2\times 2},$$

⇒ c = b/a internally stabilizes p (a = 0 ⇒ c = 1 - b IS p). • If p is internally stabilizable, then there exists $c \in K$ s.t.:

$$a = \frac{1}{1 - pc} \in A, \quad a p = \frac{p}{1 - pc} \in A, \quad b = \frac{c}{1 - pc} \in A.$$

Let $I = (a, b)$. Then, $a - b p = 1 \in I J \Rightarrow I J = A \Rightarrow I = J^{-1}$.

• Let
$$A = H_{\infty}(\mathbb{C}_+)$$
, $p = \frac{e^{-s}}{(s-1)} \in Q(A)$, $J = (1, p)$:
 $gcd\left(\frac{e^{-s}}{s+1}, \frac{s-1}{s+1}\right) = 1 \Rightarrow A : J = \{d \in A \mid d p \in A\} = \left(\frac{s-1}{s+1}\right).$

• *p* is internally stabilizable iff $\exists a, b \in A : J$ s.t. a - b p = 1:

$$\Leftrightarrow \exists x, y \in A: \begin{cases} a = \left(\frac{s-1}{s+1}\right) x, \\ b = \left(\frac{s-1}{s+1}\right) y, \\ a - b p = 1. \end{cases}$$

$$a-bp = 1 \Leftrightarrow \left(\frac{s-1}{s+1}\right) (x-py) = 1 \quad \Leftrightarrow x = \frac{s+1}{s-1} + py$$

$$\Leftrightarrow x = \frac{(s+1) + e^{-s} y}{s-1}$$

 $\Rightarrow ((s+1)+e^{-s}y(s))(1)=0 \Rightarrow y(1)=-2e.$

$$y(s) = -2 e \Rightarrow x(s) = 1 + 2 \left(\frac{1 - e^{-(s-1)}}{s-1}\right) \in A.$$

Example continued

• Therefore, we have:

$$\begin{cases} a = \left(\frac{s-1}{s+1}\right) \left(1+2\left(\frac{1-e^{-(s-1)}}{s-1}\right)\right) \in A : J, \\ b = -2e\left(\frac{s-1}{s+1}\right) \in A : J, \\ a - b p = 1. \end{cases}$$

 \Rightarrow a **stabilizing controller** *c* of *p* is defined by:

$$c = \frac{b}{a} = -\frac{2e(s-1)}{(s-1)+2(1-e^{-(s-1)})} = -\frac{2e(s-1)}{s+1-2e^{-(s-1)}}.$$

• $J = (1, p)$ is invertible, $J^{-1} = A : J = \left(\frac{s-1}{s+1}\right)$
 $\Rightarrow J = (J^{-1})^{-1} = \left(\frac{s+1}{s-1}\right)$ is principal $\Rightarrow p$ admits the coprime factorization:

$$p = \frac{n}{d}, \quad n = \frac{e^{-s}}{(s+1)}, \quad d = \frac{(s-1)}{(s+1)}, \quad dx - ny = 1.$$

SC for internal stabilizability

- Let A be an algebra of stable transfer functions and K = Q(A).
- Let $p \in K$ and J = (1, p) a fractional ideal of A.
- p admits a coprime factorization iff J is principal.
- *p* is internally stabilizable iff *J* is a invertible fractional ideal.
- If J = (k), $0 \neq k \in K$, then $J^{-1} = (1/k)$

\Rightarrow the existence of a coprime factorization is a sufficient condition for internal stabilizability.

•
$$p = n/d$$
 is a coprime factorization, $dx - ny = 1$, $x \in A$, $y \in A$,

$$\Rightarrow \begin{cases} a = dx, \\ b = dy, \end{cases} \Rightarrow c = b/a = y/x \text{ is a stabilizing controller of } p. \\ (a - bp = dx - (dy)p = dx - ny = 1, \\ a, b \in A, \quad ap = nx \in A, \quad bp = ny \in A) \end{cases}$$

Strong stabilizability

• *p* is **strongly stabilizable** iff there exists $c \in A$ such that:

$$a = \frac{1}{1 - pc} \in A$$
, $a p = \frac{p}{1 - pc} \in A$, $b = \frac{c}{1 - pc} = c a \in A$.

Using the fact that $c \in A$, we obtain:

$$J^{-1} = (a, b) = (a) = ((1 - p c)^{-1}) \Rightarrow J = (J^{-1})^{-1} = (1 - p c).$$

• We suppose that there exists $c \in A$ such that (1, p) = (1 - p c)

$$\Rightarrow \exists 0 \neq d, n \in A: \begin{cases} 1 = d(1 - pc), \\ p = n(1 - pc), \end{cases} \Rightarrow \begin{cases} p = n/d, \\ d - nc = 1, \end{cases}$$

$$\Rightarrow \left(\begin{array}{cc} 1 & -p \\ -c & 1 \end{array}\right)^{-1} = \left(\begin{array}{cc} d & n \\ d c & d \end{array}\right) \in A^{2 \times 2},$$

i.e., $c \in A$ internally stabilizes p, i.e., p is strongly stabilizable.

 \Rightarrow

• Let
$$A = H_{\infty}(\mathbb{C}_+)$$
, $K = Q(A)$, $p = \frac{(1+e^{-2s})}{(1-e^{-2s})} \in K$.

• We have
$$J = (1, p) = \left(\frac{1}{1 - e^{-2s}}\right)$$
 because:

$$\begin{cases} 1 = (1 - e^{-2s}) \frac{1}{(1 - e^{-2s})}, \\ p = \frac{(1 + e^{-2s})}{(1 - e^{-2s})} = (1 + e^{-2s}) \frac{1}{(1 - e^{-2s})}, \\ \frac{1}{(1 - e^{-2s})} = \frac{1}{2} + \frac{1}{2} \frac{(1 + e^{-2s})}{(1 - e^{-2s})}. \end{cases}$$
coprime factorization
$$\begin{cases} p = \frac{(1 + e^{-2s})}{(1 - e^{-2s})}, \\ \frac{1}{2} (1 - e^{-2s}) + \frac{1}{2} (1 + e^{-2s}) = 1. \end{cases}$$

 $\Rightarrow c = -1$ is a stable stabilizing controller of p.

• We check that $1 - pc = 1 + p = \frac{2}{(1 - e^{-2s})}$ $\Rightarrow J = (1, p) = (1/(1 - e^{-2s})) = (1 - pc).$

Robust stabilization

• $c \in K = Q(A)$ internally stabilizes $p \in K$ iff:

$$(1, p)(1, c) = (1 - p c).$$

• Let $\delta \in A$. *c* internally stabilizes *p* and *p* + δ iff we have:

$$\begin{cases} (1, p)(1, c) = (1 - pc), \\ (1, p + \delta)(1, c) = (1 - (p + \delta)c), \end{cases} \begin{cases} (1, p)(1, c) = (1 - pc), \\ (1, p)(1, c) = (1 - (p + \delta)c), \end{cases} \\ \begin{cases} (1, p)(1, c) = (1 - pc), \\ \left(\frac{1 - (p + \delta)c}{1 - pc}\right) = \left(1 - \frac{\delta c}{1 - pc}\right) = A, \end{cases} \Leftrightarrow \begin{cases} c \text{ IS } p, \\ 1 - \frac{(\delta c)}{(1 - pc)} \in U(A). \end{cases}$$

• If A is a **Banach algebra**, then (small gain theorem):

$$|1-a||_A < 1 \Rightarrow a \in U(A) = \{a \in A \mid \exists b \in A : ab = ba = 1\}$$

 \Rightarrow a sufficient condition for robust stabilization $(c/(1-p\,c)\in A)$ is:

$$\| \delta \|_{\mathcal{A}} < (\| c/(1-p c) \|_{\mathcal{A}})^{-1}.$$

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Robust stabilization

• Let $\delta \in A$. *c* internally stabilizes *p* and $p/(1 + \delta p)$ iff we have:

$$\begin{cases} (1, p) (1, c) = (1 - p c), \\ \left(1, \frac{p}{(1+\delta p)}\right) (1, c) = \left(1 - \frac{p c}{(1+\delta p)}\right), \\ \Leftrightarrow \begin{cases} (1, p) (1, c) = (1 - p c), \\ (1 + \delta p, p) (1, c) = (1 - p c + \delta p), \end{cases} \\ \Leftrightarrow \begin{cases} (1, p) (1, c) = (1 - p c), \\ \left(\frac{1 - p c + \delta p}{1 - p c}\right) = \left(1 - \frac{\delta p}{1 - p c}\right) = A, \\ \Leftrightarrow \begin{cases} c \text{ IS } p, \\ 1 - \frac{(\delta p)}{(1 - p c)} \in U(A). \end{cases} \end{cases}$$

 \Rightarrow a sufficient condition for robust stabilization $(p/(1-p\,c)\in A)$ is:

$$\|\delta\|_{A} < (\|p/(1-pc)\|_{A})^{-1}.$$

A few more results

- "IS" stands for "internally stabilized/-zable".
- "CF" stands for "coprime factorization".
- **Proposition**: Let $\delta \in A$, $p, c \in Q(A)$.
- 1. If p is **IS** by c, then p admits a **CF** \Leftrightarrow c admits a **CF**.
- 2. *p* is **IS** and *p* admits a **weakly** $CF \Leftrightarrow p$ admits a CF.
- 3. p is **IS** by $c \Leftrightarrow p + \delta$ is **IS** by $c/(1 + \delta c)$.
- 4. p is **IS** by $c \Leftrightarrow p/(1+\delta p)$ is **IS** by $c+\delta$.
- 5. p is **IS** by $c \Leftrightarrow 1/p$ is **IS** by 1/c.
- 6. *p* is externally stabilized by *c*, i.e., $pc/(1-pc) \in A$, iff:

$$(1, pc) = (1 - pc).$$

7. p = n/d CF, c = s/r CF. p is IS by $c \Leftrightarrow dr - ns \in U(A)$.

Summary

- Let A be a ring of stable transfer functions and K = Q(A).
- Let $p \in K$ be a transfer function.
- Let J = (1, p) be a fractional ideal of A and:

$$A: J = \{ d \in A \mid d \ p \in A \}.$$

- **Theorem**: 1. p is stable iff J = A iff A : J = A.
- 2. p admits a weakly coprime factorization iff:

$$\exists 0 \neq d \in A : A : J = (d).$$

Then, p = n/d, $(n = d \ p \in A)$, is a weakly coprime factorization.

3. *p* is **internally stabilizable** iff *J* is **invertible**, i.e., iff:

$$\exists a, b \in A, \quad a-b p = 1, \quad a p \in A.$$

If $a \neq 0$, then c = b/a is a **stabilizing controller** of p and:

$$J^{-1} = (a, b), \quad a = 1/(1 - pc), \quad b = c/(1 - pc).$$

Summary

4. $c \in K$ internally stabilizes $p \in K$ if we have:

$$(1, p)(1, c) = (1 - p c).$$

5. $c \in K$ externally stabilizes $p \in K$ $(p c/(1 - p c) \in A)$ iff:

$$(1, pc) = (1 - pc).$$

6. *p* is **strongly stabilizable** iff there exists $c \in A$ such that:

$$(1, p) = (1 - p c).$$

7. *p* admits a **coprime factorization** iff (1, p) is **principal**. Then, there exists $0 \neq d \in A$ such that

$$(1, p) = (1/d)$$

and p = n/d is a coprime factorization of p ($n = d \ p \in A$).

Classification of the rings A

• **Theorem**: Let A be a integral domain of stable transfer functions and K = Q(A).

1. Every transfer function $p \in K$ admits a **weakly coprime** factorization iff A is a **GCDD**, i.e., any two elements of A admits a greatest common divisor.

2. Every transfer function $p \in K$ is **internally stabilizable** iff A is a **Prüfer domain**, i.e., any f.g. ideal of A is invertible.

3. Every transfer function $p \in K$ admits a **coprime factorization** iff A is a **Bézout domain**, i.e., any f.g. ideal of A is principal.

- RH_{∞} is a **PID** \Rightarrow GCD, Prüfer and Bézout domains.
- $H_{\infty}(\mathbb{C}_+)$ is a **GCDD** but is not a Prüfer and a Bézout domain.

$$(\exists x, y \in H_{\infty}(\mathbb{C}_{+}): dx - ny = 1 \Leftrightarrow \inf_{s \in \mathbb{C}_{+}}(|d(s)| + |n(s)|) > 0$$
$$gcd(e^{-s}, 1/(s+1)) = 1, \quad \inf_{s \in \mathbb{C}_{+}}(|e^{-s}| + |1/(s+1)|) = 0.)$$
$$\widehat{\mathcal{A}} ???$$

Pre-Bézout rings

• **Definition**: An integral domain A is a **pre-Bézout ring** if, for every d, $n \in A$ such that there exists a greatest common divisor [d, n] of d and n, then there exist $x, y \in A$ satsifying:

$$dx - ny = [d, n].$$

• Example: The disc algebra $A(\mathbb{D})$ is the Banach algebra of holomophic functions in the open unit disc $\mathbb{D} = \{z \in \mathbb{C} \mid |z| < 1\}$ which are continuous on the unit circle $\mathbb{T} = \{z \in \mathbb{C} \mid |z| = 1\}$. Then, $A(\mathbb{D})$ is a **pre-Bézout ring**.

- **Proposition**: Let A be a pre-Bézout ring. Then, we have:
- 1. $p \in Q(A)$ admits a weakly coprime factorization.

2. $p \in Q(A)$ admits a coprime factorization.

Stable range

• **Definition**: A ring A has a **stable range** of A equals 1 if, for every $(d, n) \in A^{1 \times 2}$ admitting a right-inverse $(x, -y)^T \in A^2$,

i.e.,
$$dx - ny = 1$$
,

there exists $c \in A$ such that:

$$d-nc \in \mathrm{U}(A) = \{a \in A \mid \exists b \in A : ab = ba = 1\}.$$

• **Theorem**: Let A be a integral domain of transfer functions and K = Q(A). Then, every transfer function $p \in K$ which admits a coprime factorization is strongly stabilizable iff sr(A) = 1.

• Example: The following Banach algebras

 $H_{\infty}(\mathbb{D}), \quad H_{\infty}(\mathbb{C}_+), \quad A(\mathbb{D}), \quad W_+, \quad L_{\infty}(i\mathbb{R}),$

have a stable range equals to 1 (sr(RH_{∞}) = 2!)

(Treil 92, Jones/Marshall/Wolff 86, Rupp 90).

 $RH_{\infty} \subset \widehat{\mathcal{A}} \subset H_{\infty}(\mathbb{C}_+)$

"... The foregoing results about rational functions are so elegant that one can hardly resist the temptation to try to generalize them to non-rational functions. But to what class of functions? Much attention has been devoted in the engineering literature to the identification of a class that is wide enough to encompass all the functions of physical interest and yet enjoys the structural properties that allow analysis of the robust stabilisation problem",

N. Young, "Some function-theoretic issues in feedback stabilization", in *Holomorphy Spaces*, MSRI Publications 33, 1998, 337-349.

Parametrizations of all stabilizing controllers

• Theorem: Let c be a stabilizing controller of $p \in Q(A)$, a = 1/(1 - pc), b = c/(1 - pc) and J = (1, p). Then, all stabilizing controllers of p are

$$c(q_1, q_2) = rac{b + a^2 q_1 + b^2 q_2}{a + a^2 p q_1 + b^2 p q_2}, \quad (\star)$$

where q_1 and q_2 any element of A: $a + a^2 p q_1 + b^2 p q_2 \neq 0$. 1. (*) depends on **only one free parameter**

 $\Leftrightarrow p^2 \text{ admits a coprime factorization } p^2 = s/r.$ 2. If p^2 admits a coprime factorization $p^2 = s/r$,

$$(\star) \Leftrightarrow c(q) = rac{b+rq}{a+rpq}, \quad \forall q \in A: a+rpq \neq 0.$$

3. If p admits a coprime factorization p = n/d, dx - ny = 1:

$$(\star) \Leftrightarrow c(q) = \frac{y+d q}{x+n q}, \quad \forall q \in A : x+n q \neq 0.$$

K. Mori, CDC 1999, 973-975

• Let $A = \mathbb{R}[x^2, x^3]$ be the ring of discrete time delay systems without the unit delay.

• A is used for high-speed circuits, computer memory devices.

•
$$p = (1 - x^3)/(1 - x^2) \in Q(A)$$
, $J = (1, p)$.

• Using
$$(1 - x^3)(1 + x^3) = (1 - x^2)(1 + x^2 + x^4)$$
, we get

$$p = \frac{(1-x^3)}{(1-x^2)} = \frac{(1+x^2+x^4)}{(1+x^3)}$$

 $A: J = (1 - x^2, 1 + x^3)$ is not principal because $(x + 1) \notin A$.

 \Rightarrow *p* does not admit a (weakly) coprime factorization.

• As $A: J = (1 - x^2, 1 + x^3)$, we then get:

$$J(A: J) = (1 - x^2, 1 + x^3, 1 - x^3, 1 + x^2 + x^4).$$

• We have $(1 + x^3)/2 + (1 - x^3)/2 = 1 \in J(A : J)$

$$\Rightarrow \begin{cases} a = (1 + x^3)/2 \in A : J, \\ b = -(1 - x^2)/2 \in A : J, \\ a - b p = 1, \end{cases}$$

 $\Rightarrow c = b/a = -(1 - x^2)/(1 + x^3)$ internally stabilizes p.

•
$$J^{-1} = (1 - x^2, 1 + x^3) \Rightarrow J^{-2} = ((1 - x^2)^2, (1 + x^3)^2).$$

- $(x+1) \notin A \Rightarrow J^{-2}$ is not principal ideal of A.
- \Rightarrow all stabilizing controllers of *p* have the form:

$$c(q_1,q_2) = rac{-(1-x^2)+(1-x^2)^2\,q_1+(1+x^3)^2\,q_2}{(1+x^3)+(1-x^2)\,(1-x^3)\,q_1+(1+x^3)\,(1+x^2+x^4)\,q_2},$$

for all $q_1, q_2 \in A$ such that the denominator exists, $a_1, a_2 \in A$ such that the denominator exists, $a_2, a_3 \in A$

V. Anantharam, IEEE TAC 30 (1985), 1030-1031

- $A = \mathbb{Z}[i\sqrt{5}], \quad p = (1 + i\sqrt{5})/2 \in K = \mathbb{Q}(i\sqrt{5}), \quad J = (1, p).$
- Using $2 \times 3 = (1 + i\sqrt{5})(1 i\sqrt{5}) = 6$, we get

$$p = (1 + i\sqrt{5})/2 = 3/(1 - i\sqrt{5}),$$

and $A: J = (2, 1 - i\sqrt{5})$ is not a principal ideal of A.

 \Rightarrow *p* does not admit a (weakly) coprime factorization. • $J(A: J) = (2, 1 + i\sqrt{5}, 1 - i\sqrt{5}, 3) = A$ as we have $-2+3 = (-2) - (-1+i\sqrt{5}) p = 1$ $\Rightarrow c = (1 - i\sqrt{5})/2$ internally stabilizes p. • $J^{-2} = (2, 1 - i\sqrt{5})^2 = (2)$ $\Rightarrow c(q) = \frac{(1-i\sqrt{5})-2q}{2-(1+i\sqrt{5})q}, \quad \forall q \in A.$

• It is well-known that the unstable plant $p = e^{-s}/(s-1)$ is internally stabilized by the **distributed delay** controller:

$$c = -2 e (s - 1) / \left(s + 1 - 2 e^{-(s - 1)} \right).$$

$$\begin{cases}
a = \frac{1}{(1 - \rho c)} = \frac{(s + 1 - 2 e^{-(s - 1)})}{(s + 1)} \in H_{\infty}(\mathbb{C}_{+}), \\
b = \frac{c}{1 - \rho c} = -\frac{2 e (s - 1)}{(s + 1)} \in H_{\infty}(\mathbb{C}_{+}), \\
a p = \frac{p}{(1 - \rho c)} = \frac{e^{-s}}{(s + 1)} \frac{(s + 1 - 2 e^{-(s - 1)})}{(s - 1)} \in H_{\infty}(\mathbb{C}_{+}).
\end{cases}$$

• We obtain that all stabilizing controllers of p have the form:

$$c(I) = \frac{-2 \, e + I \, \frac{(s-1)}{(s+1)}}{1 + 2 \left(\frac{1-e^{-(s-1)}}{s-1}\right) + I \, \frac{e^{-s}}{(s+1)}}, I = \left(1 + 2 \, \left(\frac{1-e^{-(s-1)}}{s-1}\right)\right)^2 \, q_1 + 4 \, e^2 \, q_2.$$

• This is the **Youla-Ku**čera parametrization obtained from the following coprime factorization p = n/d:

$$n = \frac{e^{-s}}{(s+1)}, \quad d = \frac{(s-1)}{(s+1)}, \quad (-2e) n - \left(1 + 2\left(\frac{1-e^{-(s-1)}}{s-1}\right)\right) d = 1.$$

Example: Smith predictor

• Let us consider the **transfer function**:

$$p = p_0 e^{-\tau s}, \quad p_0 \in RH_{\infty}, \quad \tau \in \mathbb{R}_+.$$

• $p \in A = H_{\infty}(\mathbb{C}_+) \Rightarrow p = n/d$, where d = 1 and n = p.

 \Rightarrow the parametrization of all stabilizing controllers of p is:

$$c(q)=rac{q}{1+q\,p_0\,e^{- au\,s}}, \quad orall\, q\in A.$$

 \Rightarrow Let $c_0 \in \mathbb{R}(s)$ be a certain stabilizing controller of p_0 .

$$\Rightarrow q_{\star} = \frac{c_0}{1 - p_0 c_0} \in RH_{\infty} \subset A.$$
$$\Rightarrow c(q_{\star}) = \frac{c_0}{1 + p_0 c_0 (e^{-\tau s} - 1)} = \frac{c_0}{1 - c_0 (p_0 - p)}$$

internally stabilizes p and is called **Smith predictor**. We have:

$$\frac{p c(q_{\star})}{1 - p c(q_{\star})} = \left(\frac{p_0 c_0}{1 - p_0 c_0}\right) e^{-\tau s}.$$

Convexity of H(p, c)

- Let c be a stabilizing controller of $p \in Q(A)$.
- All stabilizing controllers of p are given by

$$egin{aligned} c(q_1,q_2) &= rac{\left(1-p\,c_*
ight)c_*+q_1+q_2\,{c_*}^2}{\left(1-p\,c_*
ight)+q_1\,p+q_2\,p\,{c_*}^2} \ orall \, q_1,\,q_2\in \mathcal{A}: \quad \left(1-p\,c_*
ight)+q_1\,p+q_2\,p\,{c_*}^2
eq 0. \end{aligned}$$

• The closed-loop system

$$\begin{pmatrix} e_1 \\ e_2 \end{pmatrix} = \begin{pmatrix} \frac{1}{1-\rho c} & \frac{p}{1-\rho c} \\ \frac{c}{1-\rho c} & \frac{1}{1-\rho c} \end{pmatrix} \begin{pmatrix} u_1 \\ u_2 \end{pmatrix} \text{ becomes :}$$

$$\begin{pmatrix} \frac{1}{1-pc_*} + q_1 \frac{p}{(1-pc_*)^2} + q_2 \frac{pc_*^2}{(1-pc_*)^2} & \frac{c_*}{1-pc_*} + q_1 \frac{1}{(1-pc_*)^2} + q_2 \frac{c_*^2}{(1-pc_*)^2} \\ \frac{p}{1-pc_*} + q_1 \frac{p^2}{(1-pc_*)^2} + q_2 \frac{(pc_*)^2}{(1-pc_*)^2} & \frac{1}{1-pc_*} + q_1 \frac{p}{(1-pc_*)^2} + q_2 \frac{pc_*^2}{(1-pc_*)^2} \end{pmatrix}.$$

$$\Rightarrow \forall \lambda \in A : \ H(p, \ c(\lambda \ q_1 + (1-\lambda) \ q_1', \ \lambda \ q_2 + (1-\lambda) \ q_2')) \\ = \lambda \ H(p, \ c(q_1, \ q_2)) + (1-\lambda) \ H(p, \ c(q_1', \ q_2')).$$

Sensitivity minimization

- Let A be a Banach algebra $(H_{\infty}(\mathbb{C}_+), \widehat{\mathcal{A}}, W_{+}...)$.
- Let c be a stabilizing controller of $p \in Q(A)$ and:

$$a = 1/(1 - p c), \quad b = c/(1 - p c) \in A.$$

• Let $w \in A$ be a weighted function. Then, we have:

 $\inf_{c \in \text{Stab}(p)} \| w/(1-p c) \|_{A} = \inf_{q_{1}, q_{2} \in A} \| w (a+a^{2} p q_{1}+b^{2} p q_{2}) \|_{A} (\star)$

 \Rightarrow (*) is now a **convex problem**.

• If p = n/d is a coprime factorization of p, dx - ny = 1,

$$\Rightarrow a + a^2 p q_1 + b^2 p q_2 = d (x + q n).$$

 $\forall \in A, \quad \exists \ q_1, \ q_2 \in A: \quad q = x^2 \ q_1 + y^2 \ q_2,$ where:

$$q_1 = d^2 (1 - 2 n y) q, \quad q_2 = n^2 (1 + 2 d x) q.$$
$$(\star) \Leftrightarrow \inf_{q \in A} \parallel w d (x + n q) \parallel_A.$$

Open questions

• What are the algebraic properties of $\widehat{\mathcal{A}}$?

• Let $\mathcal{I}(A)$ be the group of invertible fractional ideals of A and $\mathcal{P}(A)$ the group of principal fractional ideals of A.

$$\Rightarrow \mathcal{C}(A) = \mathcal{I}(A)/\mathcal{P}(A)$$

is sometimes called the **Picard group** of A. Question: $C(\hat{A})$?

- Is it possible to develop a **theory of divisors** over $H_{\infty}(\mathbb{C}_+)$?
- Let $p_1, p_2 \in K = Q(A)$. When do we have:

 $(1, p_2) \cong (1, p_1) \Leftrightarrow \exists 0 \neq k \in K : (1, p_2) = (k)(1, p_1)?$

The simultaneous stabilization problem is open when p_i ∈ Q(A), i = 1,..., n, do not admit coprime factorizations: ∃ c ∈ Q(A) : (1, p_i) (1, c) = (1 - p_i c), 1 ≤ i ≤ n?
A = {f ∈ H_∞(ℂ₊) | f(s) = f(s), ∀s ∈ ℂ₊}. Question: sr(A) = 2?

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